

End-to-End Energy Efficiency Evaluation for a Beyond 5G Wireless Communication Network

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Abstract—The energy efficiency (EE) is one of the major performance metrics for the fifth generation (5G) wireless communication system. This paper proposes an end-to-end (e2e) power consumption model and studies the energy efficiency for a heterogeneous 5G cellular architecture that can separate the indoor and outdoor communication scenarios. In this work, massive multiple-input-multiple-output (MIMO) technologies at conventional sub-6 GHz frequencies are used for long-distance outdoor communications. Light-Fidelity (LiFi) and millimeter wave (mmWave) technologies are deployed to provide a high data rate service to indoor users. Whereas, in the referenced non-separated communication system, the indoor users communicate with the outdoor massive MIMO macro base station directly. The performance of these two systems are evaluated and compared in terms of the total power consumption and energy efficiency. The results show that the separated network architecture can support a higher data rate transmission with less energy consumption, compared to the non-separate communication scenario. In addition, the results show that deploying LiFi and mmWave IAPs can enable users to transmit at a higher data rate and further reduce the energy consumption.

Keywords –Energy efficiency, 5G, massive MIMO, LiFi, mmWave.

I. INTRODUCTION

5G wireless communication systems are designed to offer a significant improvement in system capacity, spectral efficiency, average cell throughput, and EE when compared with the fourth generation (4G) wireless systems. It is widely accepted that 5G network architecture combines macrocells, picocells and small cells to support reliable, resilient and efficient wireless communications [1]. In [2], a 5G heterogeneous cellular architecture that can separate the outdoor and indoor communication scenarios is proposed. In this architecture, a macro base station (MBS) is assisted by the massive MIMO technology and antenna arrays (MBSALA) geographically distributed in the cell. Each antenna array serves a certain area and can be installed on an exterior wall or on the top of buildings which is referred to as the building mounted antenna array (BMAA). The BMAA is connected with indoor access points (IAPs) via fibres, as shown in Fig. 1. In this work, we considered two short-range IAP technologies, namely the beamforming based mmWave technology [3] and the LiFi wireless technology [4]. Since the outdoor and indoor communications operate in different frequency bands, the interference between the indoor and outdoor user equipments

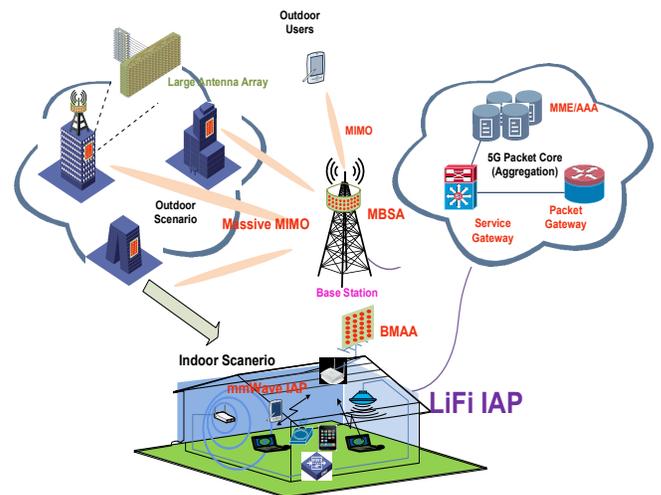


Fig. 1. A potential B5G network architecture.

(UEs) can be avoided. Besides, the high penetration loss of mmWave signals and the small coverage for the visible light signal of LiFi also reduce the interference among the IAPs deployed in the neighboring rooms and buildings. In this work, we assumed a BMAA communicates with a MBSALA using conventional Sub-6 GHz frequency band. A line of sight (LoS) path is ensured between the MBSALA and its serving BMAA. By using beamforming technologies, the LoS path can be further exploited to multiple virtual sub-beams between multiple BMAAs and a single MBSALA. As UEs can support multiple rate access technologies (RATs) in current communication systems. In this work, we assume indoor users to communicate via their small cell IAPs in the separated condition, and directly with the MBSALA in the non-separated condition. In these 5G communication scenarios, the indoor UEs can transmit at a very high data rate, of the order Gbps, with a minimum power consumption.

LiFi attocell technology modulates the data through existing illumination light emitting diode (LEDs) and reuses the illumination power to support high speed optical wireless communications [5]. The support of multiple Gbps rates with low power consumption puts LiFi-enabled wireless communication systems among the best candidate technologies for 5G mobile

wireless systems and beyond [2]. Some existing researches have focused on developing modulation techniques to increase the transmission data-rates [6] of the order Gbps at typical illumination levels in LiFi [7].

Most existing research studies to date, on EE, only focused on some specific technologies rather than a complete network, such as [8] for cognitive radio networks, [9] for massive MIMO systems, [10] for mobile femtocell systems, and [11] for relay-sided cellular networks. Referring these works, this paper develops an e2e energy consumption model for evaluating the EE of a future 5G network architecture. In this flexible network architecture, 5G networks are established with massive MIMO considering sub-6 GHz RF frequencies, small cell mmWave, and LiFi optical wireless technologies. The developed energy consumption model covers the individual aspects of a MBS, wireless communication links and small cell APs. These are relevant for the power consumption analysis, particularly the transmission bandwidth and the number of RF chains. This study can be used as a reference for the selection of communication technologies for different UEs, as well as the number of antennas in massive MIMO/LiFi communication systems. Compared to other existing works, this new model allows the detailed quantification of energy used by specific components, which enable a more accurate EE study at the network level. Offloading UEs from a small cell's IAP to the MBS and vice-versa is an important key in the 5G strategy vision to boost the capacity and EE of 5G wireless networks. Thus, whenever alternative connection technologies are available (as often happens in indoor scenarios), cellular traffic can be offloaded. The developed energy consumption model can be used to evaluate the total energy consumption of UEs in terms of their data transmission, which makes it suitable for investigating offloading cases.

The rest of this paper is organized as follows. In Sections II, channel models and signal propagation models of the proposed system are introduced. In Section III-A, the power consumption model of the 5G network architecture with mmWave IAP and LiFi attocell are explained. In Section IV, the results of performance evaluation for both scenarios are discussed. Finally, Section V draws conclusions of this paper.

II. CHANNEL MODEL AND SIGNAL DETECTION

An outdoor massive MIMO channel which is represented by a beamforming channel ℓ at MBSALA i and BMAA j , can be mathematically described by

$$\mathbf{G}_{\ell ji} = \sqrt{\beta_{\ell ji} M_T M_R} \mathbf{a}_{M_R}^*(\phi_{\ell ji}) \mathbf{a}_{M_T}^T(\theta_{\ell ji}) \quad (1)$$

where $j = \sqrt{-1}$, $\theta_{\ell ji}$ is the angle of arrival (AoA) at MBSALA i , M_T and M_R are the number of transmission and receiver antenna elements, respectively, $\phi_{\ell ji}$ is the angle of arrival (AoA) at BMAA j , $\beta_{\ell ji}$ is the path loss, and

$$\mathbf{a}_{M_R/T}(\theta) = \frac{1}{\sqrt{M_{R/T}}} [1, \dots, e^{-j2\pi\Delta(m-1)\theta}, e^{-j2\pi\Delta(M_{R/T}-1)\theta}]^T \quad (2)$$

is the antenna response of receiver or transmitter side. Δ is the normalized antenna separation. Without loss of generality, we assume $\Delta = 1/2$. The transmission beam vector at BMAA j is $\mathbf{a}_{M_T}^*(\theta_{\ell ji})$, and the received beam vector at BMAA j is $\mathbf{a}_{M_R}^T(\phi_{\ell ji})$. $\theta_{\ell ji}$ and $\phi_{\ell ji}$ can be adjusted and optimized when

deploying MBSALA and BMAA, and thus the beamforming channel between MBSALA and BMAA can be considered static and known to both sides.

The downlink signal vector of MBSALA i , \mathbf{x}_i , is a linear combination of beamformed signals destined to the N_b BMAs, which can be expressed as

$$\mathbf{x}_i = \sum_{j=1}^{N_b} \sum_{\ell=1}^L \mathbf{b}_{\ell ji} s_{\ell ji} \quad (3)$$

where $s_{\ell ji}$ is an i.i.d. complex Gaussian random variable with zero mean and variance $P_{\ell ji}$, $\mathbf{b}_{\ell ji} = \mathbf{a}^*(\theta_{\ell ji})$ denotes the beamforming vector ℓ to BMAA j . In this work, the maximum-ratio combining (MRC) precoding is applied. A perfect channel state information (CSI) is known at the MBSALA.

The received signal at BMAA j over the beamforming channels ℓ is given by

$$r_{\ell ji} = \mathbf{a}^T(\phi_{\ell ji}) (\mathbf{G}_{\ell ji} \mathbf{x}_i + \mathbf{n}) = \sqrt{\beta_{\ell ji} M_T M_R} s_{\ell ji} + n' \quad (4)$$

where $\mathbf{a}^T(\phi_{\ell ji})$ is the received beam vector ℓ at BMAA j , \mathbf{n} is the received Gaussian noise vector with $\mathbf{n} \sim \mathcal{CN}(\mathbf{0}, \sigma_n^2 \mathbf{I}_{M_R})$. As the link between the BMAA and MBSALA is relative stable, the orthogonality of transmitted and received beamforming vectors enables the receivers to avoid any interference from signals sent to other buildings.

For the indoor mmWave link, the expression of channel model is very similar, but the interference from other indoor users should be considered. The received signal at the k -th UE can be given by

$$r_k = \mathbf{g}_k^T \mathbf{b}_{m_k} s_k + \sum_{j=1, j \neq k}^{N_{iue}} \mathbf{g}_k^T \mathbf{b}_{m_j} s_j + n \quad (5)$$

$$= \sqrt{\beta_k} F_{M_T}(\theta_{m_k} - \theta_k) + \sqrt{\beta_k} \sum_{j=1, j \neq k}^{N_{iue}} F_{M_T}(\theta_{m_k} - \theta_j) s_j + n$$

where $F_M(x) = \frac{\sin \pi M x / 2}{M \sin(\pi x / 2)}$ is the Sinc function. $\theta_{m_k} - \theta_k$ is the difference between the beam vector and the angle of departure (AoD).

For the indoor LiFi link, a desirable LoS channel gain of LiFi attocell IAP is given by

$$H_{\text{LOS}} = \begin{cases} \frac{(m+1)A}{2\pi d^2} \cos^m \phi g_f g(\psi) \cos \psi, & 0 \leq \psi \leq \Psi_c \\ 0, & \psi_{ij} > \Psi_c \end{cases} \quad (6)$$

where A denotes the physical area of detector, d denotes the distance between the LiFi attocell IAP and receivers (UEs), ϕ and ψ denote the angle of radiance with respect to the axis normal to the transmitter surface, and the angle of incidence with respect to the axis normal to the receiver surface, respectively. g_f denotes the gain of optical filter, and Ψ_c denotes the field-of-view (FoV) of receiver. Note that $g(\psi) = \zeta^2 / \sin^2 \Psi_c$, for $0 \leq \psi_i \leq \Psi_c$, and 0 for $\psi_i > \Psi_c$, represents the optical concentrator gain, where ζ denotes the refractive index. The Lambertian order is obtained from $m = -1 / \log_2(\cos \Phi_{1/2})$, where $\Phi_{1/2}$ is the half-intensity angle [12].

The radiance angle, ϕ , and the incidence angle, ψ , of the LiFi IAP and the receiver are given based on the analytical geometry rules, such as $\cos \phi = \mathbf{d} \cdot \mathbf{n}_{\text{tx}} / \|\mathbf{d}\|$,

and $\cos \psi = -\mathbf{d} \cdot \mathbf{n}_{\text{rx}} / \|\mathbf{d}\|$, where $\mathbf{n}_{\text{tx}} = [0, 0, -1]$ and $\mathbf{n}_{\text{rx}} = [0, 0, 1]$ are the normal vectors at the LiFi IAP and the receiver planes, respectively. And \mathbf{d} denotes the distance vector between the LiFi AP and the receiver, and $\|\cdot\|$ and “ \cdot ” stand for the Euclidean norm operators and the inner product, respectively [12].

For more information about 5G channel models, please refer to [13], [14], [15].

III. ENERGY EFFICIENCY

Energy efficiency, η , measures the effectiveness of converting power into data traffic transmission. It is defined as the spectral efficiency divided by the total power consumption, given by

$$\eta = \frac{\mathbb{C}}{\overline{P}}. \quad (7)$$

The spectral efficiency is defined by the Shannon equation as

$$\mathbb{C} = \gamma \mathbb{E}[\log_2(1 + \text{SINR})] \quad (8)$$

where SINR is the signal-to-interference-noise-ratio under the given channel, and γ is the channel usage efficiency factor. Based on the approximation derived in [16], for a massive MIMO system, we assume $\mathbb{E}[\log_2(1 + \text{SINR})] \approx \log_2(1 + \mathbb{E}[\text{SINR}])$, the spectral efficiency analysis can be reduced to the analysis of the expectation of SNR, $\mathbb{E}[\text{SNR}]$.

Based on Eq.(4), the expectation of received SNR over a link ℓ between the MBSALA i and BMAA j can be expressed as

$$\mathbb{E}[\text{SNR}_{\ell ji}^{\text{M} \rightarrow \text{B}}] = \frac{\beta_{\ell ji} M_T P_{\ell ji}}{\sigma_n^2 / M_R} \quad (9)$$

where $P_{\ell ji}$ denotes the signal variance. Similar to 9 For the indoor mmWave link, its SINR can be expressed as

$$\mathbb{E}[\text{SINR}_k^{\text{I} \rightarrow \text{U}}] = \frac{\beta_k \mathbb{E}[F_M^2(\theta_k - \theta_{m_k})] P_k}{\beta_k \mathbb{E}[\sum_{j=1, j \neq k}^{N_{\text{IUE}}} F_M^2(\theta_k - \theta_{m_j})] P_j + \sigma_n^2}. \quad (10)$$

For the LiFi link, the SINR is given by:

$$\text{SINR} = \frac{c_f^2 P_t^2 H_{\text{LOS}}^2}{N_0 B + \sum_{j=1}^{N_F} c_{Ij}^2 P_{Ij}^2 H_{Ij\text{LOS}}^2} \quad (11)$$

where N_F denotes the number of interfering LiFi attocell IAPs, I denotes the symbol for interfering LiFi attocell IAPs, c_f denotes the LED coefficient [17], P_t denotes the optical transmitted power, N_0 denotes the noise spectral density.

A. System power model

The total power consumption of the system is the sum of the power consumed by all the power devices deployed in a wireless communication cell. In the separate outdoor and indoor scenario, the total power consumed by MBS, MBSALA, BMAA and IAP can be expressed as follows:

$$P_{\text{cell}} = P_{\text{tot}}^{\text{MBS}} + \sum_{i=1}^{N_a} (P_{\text{tot}}^{\text{BMAA}} + P_{\text{tot}}^{\text{IAP}}). \quad (12)$$

where N_a denotes the number of MBSALA. A high level energy efficiency evaluation framework (E³F) for mobile communication systems was investigated and developed by the Energy Aware Radio and network technologies (EARTH) project [18]. Based on the framework, the general energy consumption model of a mobile communication system is given by

$$P_{\text{sys}} = P_{\text{BB}} + P_{\text{RF}} + P_{\text{PA}} + P_{\text{OH}} \quad (13)$$

where P_{BB} represents the power consumed by the digital baseband processing unit, P_{RF} and P_{RF} represent the power consumption of RF front-end (FE) and power amplifier (PA), and P_{OH} represents the power overhead. This is mainly consumed by the system cooling unit and AC-DC converters.

The digital baseband processing includes digital signal processing, and system control and network processing. For digital signal processing, the operations include digital filtering, up/down sampling, (I)FFT, MIMO channel training and estimation, OFDM modulation/demodulation, symbol mapping and channel encoding/decoding. The operation complexity denoted by \mathcal{O} , is measured by Giga floating-point operations per second (GFO/S), depending on the type of the operation and number of UEs and data streams. The power consumption per Giga floating-point operation is further scaled by a technology-dependent factor ρ . Thus, P_{BB} can be obtained by dividing GFO/S by ρ . For systems nowadays, the factor is $\rho = 160$ GOP/W.

The key constituent components of RF FE include carrier modulators, frequency synthesis, clock generators, digital to analogue /analogue to digital converters, mixers and so on. The power consumption of these components scale with parameters, namely system bandwidth, number of antennas and traffic load. The power consumption model of PA depends on the type and the maximum output power of the amplifier. It is also related to the actual output power that assures the desired spectral efficiency. This paper considers two types of power amplifiers: Class-B PA and Doherty PA. The class-B PA is equipped in the MBSALA and BMAA, which enables relatively high output power transmission. The Doherty PA is developed for high-frequency band communication systems with high power efficiency. It is deployed in the IAPs. The power models for MBS, MBSALA, BMAA and IAP are developed in the following subsections.

1) *Power model of MBS*: The total power of MBS is given by

$$P_{\text{tot}}^{\text{MBS}} = \frac{P_{\text{BB}}^{\text{MBS}} + N_a P_{\text{tot}}^{\text{MBSALA}}}{(1 - \eta_c)(1 - \eta_{\text{ac-dc}})(1 - \eta_{\text{dc-dc}})} \quad (14)$$

where $P_{\text{BB}}^{\text{MBS}}$ is the power of digital base band processing at MBS, $P_{\text{tot}}^{\text{MBSALA}}$ is the total power consumed by a MBSALA, and η_c , $\eta_{\text{ac-dc}}$, $\eta_{\text{dc-dc}}$ are the power efficiency of the cooling system, AC-DC and DC-DC conversion, respectively.

Based on the outdoor distributed antenna architecture, the digital baseband processing at MBS includes mapping/demapping of symbols, channel encoding (\mathcal{O}_{enc}), upper layer network (\mathcal{O}_{nw}) and control operations ($\mathcal{O}_{\text{ctrl}}$) for the N_a MBSALA, respectively. Thus, $P_{\text{BB}}^{\text{MBS}}$ can be further written by

$$P_{\text{BB}}^{\text{MBS}} = \sum_{i=1}^{N_a} (\mathcal{O}_{\text{ctrl},i} + \mathcal{O}_{\text{network},i} + \mathcal{O}_{\text{enc},i}) / \rho. \quad (15)$$

2) *Power model of MBSALA*: The power consumption of MBSALA can be decomposed as

$$P_{\text{tot}}^{\text{MBSALA}} = P_{\text{BB}}^{\text{MBSALA}} + P_{\text{RF}}^{\text{MBSALA}} + P_{\text{PA}}^{\text{MBSALA}}. \quad (16)$$

For MBSALA, the downlink baseband processes include filtering, up sampling, (I)FFT of OFDM symbols, massive MIMO channel estimation, precoding/beamforming, symbol mapping, and control and network relation operations, which can be expressed as

$$P_{\text{BB}}^{\text{MBSALA}} = (\mathcal{O}_{\text{fltr}} + \mathcal{O}_{\text{fft}} + \mathcal{O}_{\text{est}} + \mathcal{O}_{\text{bf}} + \mathcal{O}_{\text{pre}} + \mathcal{O}_{\text{map}} + \mathcal{O}_{\text{ctrl}} + \mathcal{O}_{\text{nw}}) / \rho. \quad (17)$$

Here the complexity of (I)FFT can be scaled by $\mathcal{O}_{(i)\text{fft},i} = N_s N_{(i)\text{fft}} \log_2(N_{(i)\text{fft}})$, where $N_{(i)\text{fft}}$ is the number of sub-carriers of an OFDM symbol and N_s is the total number of OFDM symbols. The complexity of channel estimation by correlation of orthogonal pilot sequences can be scaled by $\mathcal{O}_{\text{est}} = \tau M_T N_{\text{ue}}$, where N_{ue} denotes the number of UEs, and we let $\tau = N_{\text{ue}}$. The complexity of channel precoding and beamforming operations can be scaled by $\mathcal{O}_{\text{bf/pre}} = (N_{\text{ue}} + N_b L)(1 - \tau/N_c)$ for uplink channel estimation. The RF FE (front-end) of MBSALA has modulator, mixer, clock generation and D-A converter. The total power consumption of RF FE (front-end) is given by

$$P_{\text{RF}}^{\text{MBSALA}} = M_T (P_{\text{mod}}^{\text{MBSALA}} + P_{\text{mix}}^{\text{MBSALA}} + P_{\text{dac}}^{\text{MBSALA}}) + \sqrt{M_T} P_{\text{clk}}^{\text{MBSALA}}. \quad (18)$$

The power consumption of the modulator, mixer and DAC scales linearly with the number of antennas. The power consumption of clock generator scales by the square root of the number of antennas.

3) *Power consumption model of BMAA*: The power consumption model of BMAA can be expressed as follows:

$$P^{\text{BMAA}} = \frac{P_{\text{BB},\ell}^{\text{BMAA}} + M_R P_{\text{RF}}^{\text{BMAA}}}{(1 - \eta_c)(1 - \eta_{\text{ac-dc}})(1 - \eta_{\text{dc-dc}})}. \quad (19)$$

In this work, we assume a BMAA receives the downlink traffic from its serving MBSALA over L beamforming data links and forwards it to an IAP. The baseband data processes consist of filtering, beamforming process, sampling, IFFT process, symbol de-mapping, channel decoding, control and network processes. The GOP/S can be estimated similarly to the case of MBSALA. The power consumption of baseband processing in BMAA can be given as follows:

$$P_{\text{BB}}^{\text{BMAA}} = L \cdot (\mathcal{O}_{\text{fltr}} + \mathcal{O}_{\text{bf}} + \mathcal{O}_{\text{smp}} + \mathcal{O}_{\text{ifft}} + \mathcal{O}_{\text{demap}} + \mathcal{O}_{\text{dec}} + \mathcal{O}_{\text{ctrl}} + \mathcal{O}_{\text{nw}}) / \rho. \quad (20)$$

The analogue components of the downlink RF FE include mixer, clock, variable gain amplifiers (VGA), ADC and low-noise amplifier (LNA). The power consumption of RF in BMAA can be given as follows:

$$P_{\text{RF}}^{\text{BMAA}} = M_R \cdot (\mathcal{O}_{\text{mix}} + \mathcal{O}_{\text{vga}} + \mathcal{O}_{\text{adc}} + \mathcal{O}_{\text{lna}}) + \sqrt{M_R} \mathcal{O}_{\text{clk}}. \quad (21)$$

Note that the power consumption of LNA and VGA over the receiver RF FE are considered constant, unlike the PA used for signal transmission.

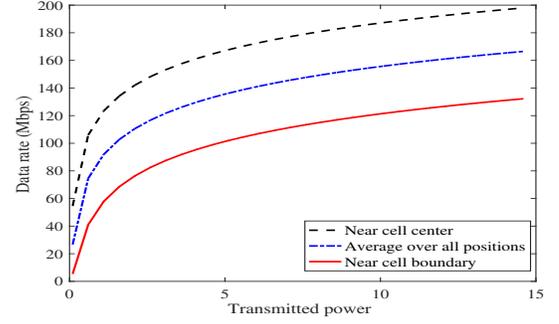


Fig. 2. LiFi Downlink throughput distribution Vs. LiFi IAP transmitted power.

4) *Power consumption Model of mmWave IAP*: For the mmWave IAP, its power consumption model can be expressed as

$$P_{\text{tot}}^{\text{IAP}} = \frac{P_{\text{BB}}^{\text{IAP}} + M'_T (P_{\text{RF}}^{\text{IAP}} + P_{\text{PA}}^{\text{IAP}})}{(1 - \eta_c)(1 - \eta_{\text{ac-dc}})(1 - \eta_{\text{dc-dc}})} \quad (22)$$

where

$$P_{\text{RF}}^{\text{IAP}} = M'_T (P_{\text{mix}}^{\text{IAP}} + P_{\text{dac}}^{\text{IAP}} + P_{\text{bft}}^{\text{IAP}} + P_{\text{fs}}^{\text{IAP}}) + \sqrt{M'_T} P_{\text{clk}}^{\text{IAP}} \quad (23)$$

and

$$P_{\text{PA}}^{\text{IAP}} = \begin{cases} \frac{2}{\pi} \sqrt{P_o P_{\text{max}}^{\text{IAP}}}, & 0 < P_o < 0.25 P_{\text{max}}^{\text{IAP}} \\ \frac{6}{\pi} \sqrt{P_o P_{\text{max}}^{\text{IAP}}}, & 0.25 P_{\text{max}}^{\text{IAP}} \leq P_o \leq P_{\text{max}}^{\text{IAP}} \end{cases}. \quad (24)$$

5) *Power consumption Model of LiFi IAP*: A LiFi attocellular network consists of several small attocells. Each IAP covers an area with radius 3 – 10 m. The power consumed for lighting the off-the-shelf LEDs in the IAP is used to support visible light communication [4]. The light photons are modulated at a very high speed to support 3.5Gbps @2m distance as well as 1.11Gbps @10m, with a total optical output power of 5 mW [19], [20]. This provides a significant spectrum, which can support hungry bandwidth applications and emerging services with low power consumption.

The maximum number of transmitted bits per joule of input energy in a LiFi communication system is known as the energy consumption factor (CF) [21]. The total power consumption in the LiFi attocellular system comprises two main parts: the circuit (illumination) power consumption and the power consumed for transmitting the wireless data at high data rates [17]. The illumination power can be expressed as follows [17]:

$$P_{\text{Light}} = \frac{nqV_T \Phi}{p_f \epsilon} \ln\left(\frac{q\Phi}{p_f \epsilon I_s} + 1\right); \quad (25)$$

And the extra power consumed for transmission high data rates can be expressed as follows [17]:

$$P_{\text{Comm}} = \frac{nqV_T H_{\text{LOS}}^2}{2p_f \epsilon \mu_\Phi}; \quad (26)$$

where the variables in Eq. (25) and Eq. (26) are defined in Eqs.(3,5) in [17]. Fig. 2 shows the throughput distribution vs. the transmitted power across a LiFi attocell.

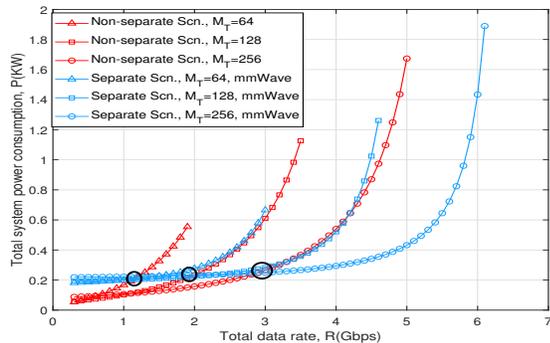


Fig. 3. Comparison of system power consumption for non-separate and separate scenarios with mmWave IAPs.

IV. PERFORMANCE RESULTS

The performance of the proposed 5G network architecture is evaluated and compared using MATLAB in terms of total power consumption and energy efficiency for LiFi and mmWave IAPs. The performance evaluation parameters for the mmWave AP are based on the reference [3]. It is considered that each MBSALA serves 4 buildings. Each MBSALA is equipped with a number of antennas, which is taken to be (64, 128, 256) in the separate and non-separate evaluation scenarios. For each BMAA, 64 antennas are equipped. For each MBSALA and BMAA pair, 4 beamforming links are established. The WINNER II [22] B5a Rooftop (Eq.(4.23), Page 43)- Rooftop model is adopted as the path loss model for the link between the MBSALA and BMAA, which is given by

$$PL_{MBSALA} = 23.5 \log_{10}(d) + 42.5 + 20 \log_{10}(f_c/5.0). \quad (27)$$

For the non-separate coverage system, the indoor users are served directly by the MBSA and a penetration loss of 20 dB is considered.

Simulation results in Fig. 3 shows the total end-to-end power consumption of the total data rate transmission in separate and non-separate scenarios. A mmWave IAP is considered with the configuration parameters used in [3]. From this figure, it can be observed that the increase in the number of antennas, M_T , provides an opportunity to support more users and hence data rate transmission. The total power consumption increases in terms of data rate transmission in both separate and non-separate scenarios. However, at relatively low data rate transmission, the total power consumption in separate and non-separate scenarios remains steady, despite the increase of M_T . For example, the total power consumption in the non-separate scenario remains steady until 1 Gbps, when M_T increases from 64 to 128. Similarly, the total power consumption in the separate scenario remains steady until 3 Gbps, when M_T increases from 64 to 128. This can be attributed to that fact that when M_T increases, the antennas and IAPs are more likely to become saturated. For a given M_T , it is observed that the power consumption curves of the two scenarios cross at some point. On the right side of the crossing point, the total power consumption in the separate scenario consumes less power than in the non-separate scenario to transmit the same data rate. But on the left side of the crossing point, this trend is in the opposite direction. Fig. 3 shows that more total power

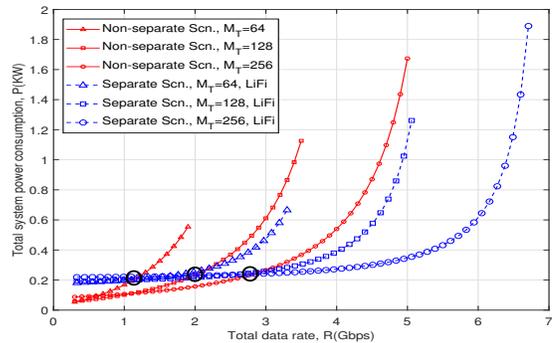


Fig. 4. Comparison of system power consumption for non-separate and separate scenarios with LiFi IAPs.

is consumed in the separate scenario than in the non-separate scenario at a low data rate. But, the results show noticeable opposite trends when the data rate is significantly increased. This is attributed to the BMAA and IAPs deployed in the separate scenario, which consume less power to transmit a high data rates. For example, when $M_T=256$, data rate= 5 Gbps, the total power consumption of the non-separate scenarios is 1.6 KW, while for the separated case, its consumed power is only 0.4 KW. In this case, it is recommended that the non-separate scenario is deployed. However, an IAP LiFi attocell that serves indoor users has reduced the total energy consumption at low and high data rates transmission by almost 10% compared with a mmWave IAP, as shown in Fig. 4. The separate scenario supports great data rate transmission with less power consumption, particularly in the case of deploying LiFi IAPs for serving UEs, as shown in Fig. 3 and Fig. 4. Nevertheless, all the result trends show an increase in the power consumption when the total transmitted data rate exceeds a specific threshold value. This is because the power amplifiers reach their saturation point after the total power consumption exceeds a certain level.

The EE versus SE trade-off curves of the two scenarios are shown in Fig. 5 and Fig. 6. The trends of EE firstly ascend as SE increases and start to descend after a certain apex point. For EE, it shows that the increase in the number of antennas do not change the energy efficiency at low SE due to the extra power consumed by BMAA and IAP. However, the EE increases sharply at high SE when the number of antennas increases. From these figures, it is clear that by separating the indoor and outdoor communication scenarios, the system can obtain a better trade-off between SE and EE. This demonstrates, in turn, the advantage of the proposed 5G network and communication architecture.

V. CONCLUSIONS

This paper has introduced an e2e power consumption model for a 5G network architecture which integrated massive MIMO, indoor mmWave, and LiFi attocell technologies. The obtained results show that by separating the indoor and outdoor communication scenarios, the 5G system can obtain a better trade-off between EE and SE. This indicates the advantage of the proposed 5G network architecture. The obtained results also indicates that the indoor UEs may be need to communicate directly to the outdoor MBS when their transmit data rate is relatively low, to achieve a better EE performance. In contrast, the UEs can communicate via the LiFi and mmWave

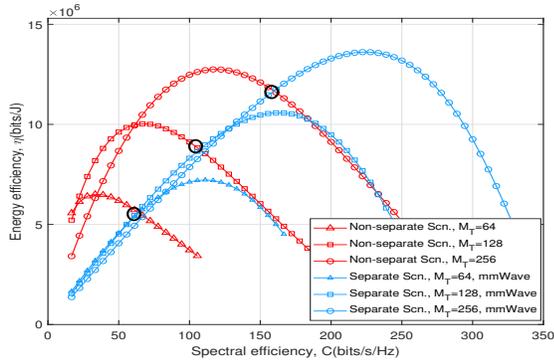


Fig. 5. EE and SE trade off curves for separate and non-separate scenarios with mmWave IAPs.

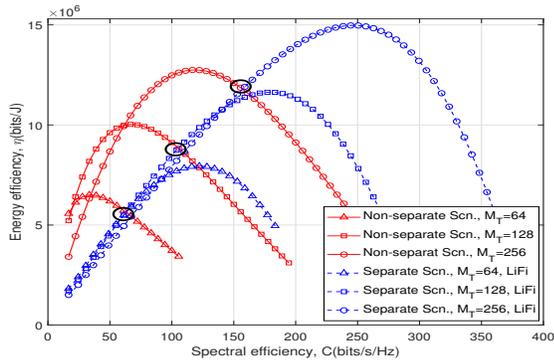


Fig. 6. EE and SE trade off curves for separate and non-separate scenarios with LiFi IAPs.

IAPs, when their data rate is high. The general trends of performance metrics in the obtained results also confirm the performance requirement of 5G to support greater data rates, in the order of multiple Gbps, with low power consumption and higher energy efficiency. This means that the UEs may require some metric indicators to guide them to communicate via different technologies. This can be a design issue for controlling traffic offloading from cellular networks to small cell IAPs and vice-versa. The results also show that deploying LiFi attocell IAPs can reduce the total power consumption by almost 10% compared to the RF mmWave indoor wireless small cell technology, which shows the potential advance of LiFi.

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